



UCC36351

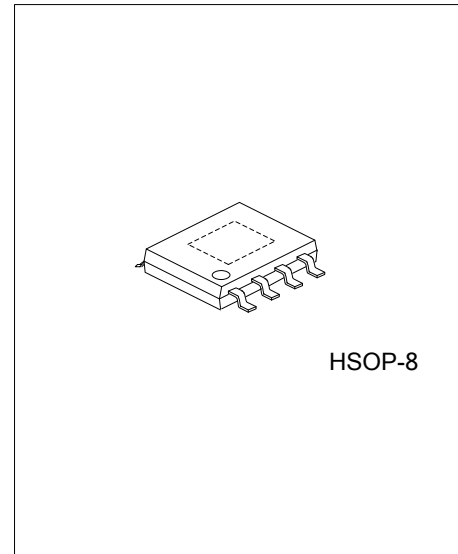
Preliminary

CMOS IC

36V SYNCHRONOUS BUCK CONVERTER WITH CC/CV

DESCRIPTION

UTC **UCC36351** is a wide input voltage, high efficiency Active CC step-down DC/DC converter that operates in either CV (Constant Output Voltage) mode or CC (Constant Output Current) mode. UTC **UCC36351** provides up to 3.5A output current at 160kHz switching frequency. Current mode control provides fast transient response and cycle-by-cycle current limit. An internal soft-start prevents inrush current at turn-on.



FEATURES

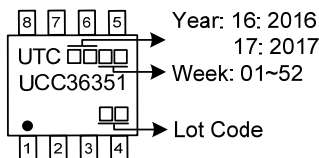
- * Wide 8V~36V Operating Input Range
- * Integrated 75/45mΩ Power MOSFET Switches
- * Output Adjustable from V_{FB} (1.00V±2%) to 16V
- * Up to 93% Efficiency
- * Internal Soft-Start and Fixed 160KHz Frequency
- * Stable with Low ESR Ceramic Output Capacitors
- * Cycle-by-Cycle Over Current Protection
- * Input Under/Over Voltage Lockout

ORDERING INFORMATION

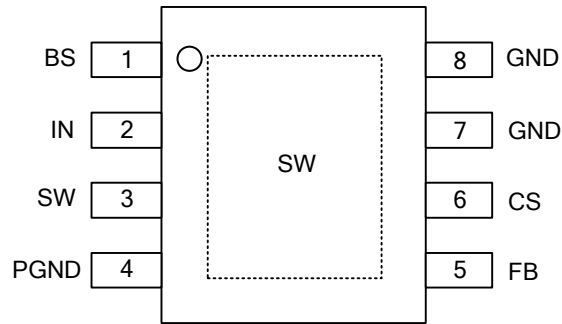
Ordering Number		Package	Packing
Lead Free	Halogen Free		
UCC36351L-SH2-R	UCC36351G-SH2-R	HSOP-8	Tape Reel

<p>UCC36351G-SH2-R</p> <p>(1) Packing Type (2) Package Type (3) Green Package</p>	<p>(1) R: Tape Reel (2) SH2: HSOP-8 (3) G: Halogen Free and Lead Free, L: Lead Free</p>
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MARKING



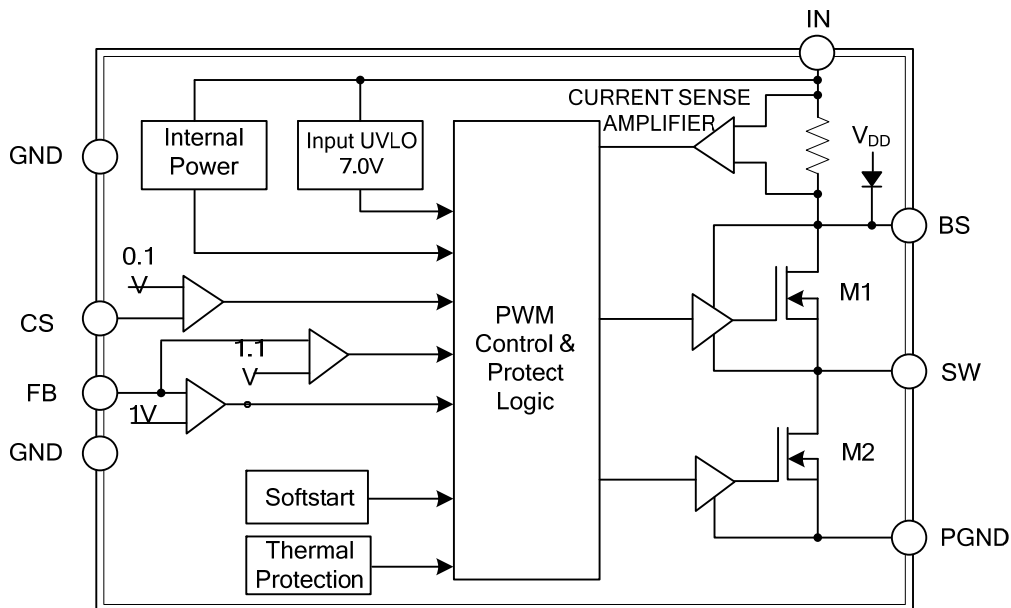
■ PIN CONFIGURATION



■ PIN DESCRIPTION

PIN NO.	PIN NAME	DESCRIPTION
1	BS	Boot-Strap Pin. Supply high side gate driver. Decouple this pin to LX pin with 24ohm+0.1uF ceramic cap.
2	IN	Power Input pin. Bypass IN to GND with a suitably large capacitor to eliminate noise on the input to the IC.
3	SW	Power Switching Output. SW is the switching node that supplies power to the output. Connect the output LC filter from SW to the output load.
4	PGND	Power Ground.
5	FB	Feedback Input. FB senses the output voltage to regulate that voltage. Drive FB with a resistive voltage divider from the output voltage.
6	CS	The Current Sense pin.
7, 8	GND	Ground.

■ BLOCK DIAGRAM



■ ABSOLUTE MAXIMUM RATING ($T_A=25^{\circ}\text{C}$, unless otherwise specified)

PARAMETER	SYMBOL	RATINGS	UNIT
Supply Voltage	V_{IN}	-0.3 ~ +42	V
Output Voltage	V_{OUT}	$V_{FB} \sim 6$	V
Switch Node Voltage	V_{SW}	-0.3 ~ $V_{IN}+0.3$	V
Boost Voltage	V_{BS}	$V_{SW}-0.3 \sim V_{SW}+6$	V
All Other Pins		-0.3 ~ +6	V
Junction Temperature	T_J	150	$^{\circ}\text{C}$
Storage Temperature	T_{STG}	-65 ~ +150	$^{\circ}\text{C}$

Note: Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

■ RECOMMENDED OPERATING RATINGS

PARAMETER	SYMBOL	RATINGS	UNIT
Ambient Operating Temperature	T_A	-40 ~ +85	$^{\circ}\text{C}$

■ THERMAL CHARACTERISTICS

PARAMETER	SYMBOL	RATINGS	UNIT
Junction to Ambient	θ_{JA}	40	$^{\circ}\text{C}/\text{W}$
Junction to Case	θ_{JC}	15	$^{\circ}\text{C}/\text{W}$

Note: θ_{JA} is measured with the PCB copper area of approximately 1 in^2 (Multi-layer). That need connect to exposed pad.

■ ELECTRICAL CHARACTERISTICS ($V_{IN}=12\text{V}$, $T_A=25^{\circ}\text{C}$, unless otherwise specified)

PARAMETER	SYMBOL	TEST CONDITIONS	MIN	TYP	MAX	UNIT
Input Voltage Range			8		36	V
Quiescent Current	I_{CCQ}	$V_{FB}=1.05\text{V}$		1	1.5	mA
Feedback Voltage	V_{FB}	$8\text{V} \leq V_{IN} \leq 36\text{V}$	0.98	1.00	1.02	V
Feedback Overvoltage Threshold	$\text{OVP}_{(FB)}$			1.1X		V_{FB}
Cable Compensation Current (Note)	I_{CFB}	$V_{CS}=100\text{mV}$		4		μA
High-Side Switch On Resistance (Note)	$R_{DS(ON)1}$			75		$\text{m}\Omega$
Low-Side Switch On Resistance (Note)	$R_{DS(ON)2}$			45		$\text{m}\Omega$
High-Side Switch Leakage Current		$V_{FB}=1.05\text{V}$, $V_{SW}=0\text{V}$			10	μA
Upper Switch Current Limit		Minimum Duty Cycle	4.0	5.0		A
Oscillation Frequency	F_{OSC1}			160		KHz
Short Circuit Oscillation Frequency	F_{OSC2}	$V_{FB} < 0.4\text{V}$		80		KHz
Short Circuit Retry time (Note)	RT_{SCP}	$V_{FB} < 0.5\text{V}$		1		mS
Maximum Duty Cycle	D_{MAX}			88		%
Minimum On Time (Note)	$T_{ON(min)}$			220		ns
Current Sense Voltage	V_{CS}		90	100	110	mV
V_{IN} OVP Turn-Off Voltage		Input Voltage Rising		36		V
V_{IN} OVP Hysteresis		Input Voltage Falling		3		V
Input Under Voltage Lockout Threshold	UVLO	V_{IN} Rising	6.5	7.0	7.5	V
Input Under Voltage Lockout Threshold Hysteresis	UVLO-Hys			800		mV
Soft-Start Period				2		ms
Thermal Shutdown	T_{SD}			155		$^{\circ}\text{C}$
Thermal Shutdown Hysteresis	T_{SH}			30		$^{\circ}\text{C}$

Note: Guaranteed by design.

■ FUNCTION DESCRIPTIONS

The UTC **UCC36351** is a synchronous rectified, current-mode, step-down regulator. It regulates input voltages from 8V to 36V down to an output voltage as low as V_{FB} , and supplies up to 3.5A of load current.

The UTC **UCC36351** uses current-mode control to regulate the output voltage. The output voltage is measured at FB through a resistive voltage divider and amplified through the internal Transconductance error amplifier.

The converter uses internal N-Channel MOSFET switches to step-down the input voltage to the regulated output voltage. Since the high side MOSFET requires a gate voltage greater than the input voltage, a boost capacitor connected between SW and BS is needed to drive the high side gate. The boost capacitor is charged from the internal 5V rail when SW is low.

When the UTC **UCC36351** FB pin exceeds 10% of the nominal regulation voltage of V_{FB} , the over voltage comparator is tripped and the COMP pin is discharged to GND, forcing the high-side switch off.

■ APPLICATION INFORMATION

Setting the Output Voltage

The output voltage is set using a resistive voltage divider from the output voltage to FB pin. The voltage divider divides the output voltage down to the feedback voltage by the ratio.

Thus the output voltage is:

$$V_{OUT} = V_{FB} \times \frac{R1 + R2}{R2}$$

For example, $V_{FB}=1.00V$ for a 5.0V output voltage, $R2$ is 10k Ω , and $R1$ is 40k Ω .

Inductor Selection

The inductor is required to supply constant current to the output load while being driven by the switched input voltage. A larger value inductor will result in less ripple current that will result in lower output ripple voltage. However, the larger value inductor will have a larger physical size, higher series resistance, and/or lower saturation current. A good rule for determining the inductance to use is to allow the peak-to-peak ripple current in the inductor to be approximately 30% of the maximum switch current limit.

VIN	<28V	<35V
Inductor	33uH	33uH

The choice of which style inductor to use mainly depends on the price vs. size requirements and any EMI requirements.

Output Short-Circuit protection

The UTC **UCC36351** provides output short-circuit protection retry function. When V_{OUT} is short ($V_{FB}<0.5V$), the auto restart function can be started that restart the regulator cycle by cycle.

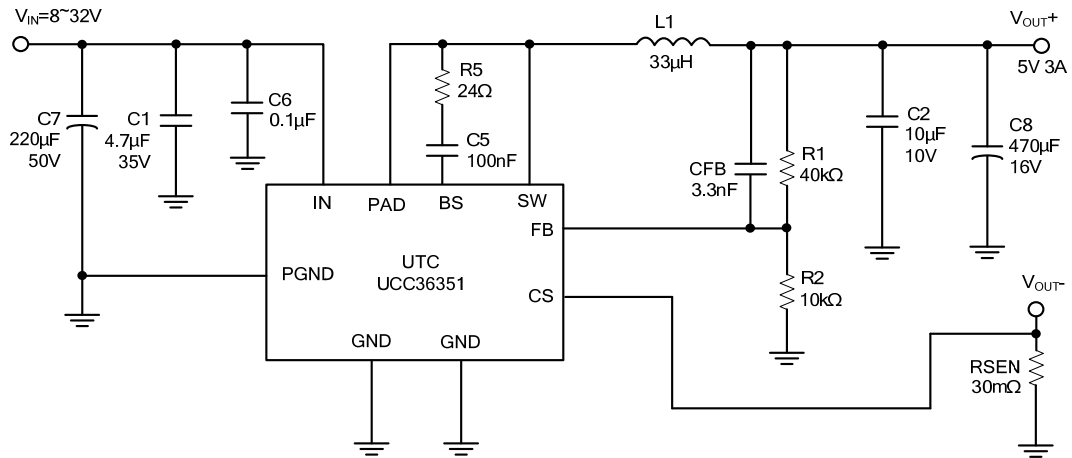
(Retry time 2mS, Shutdown regulator time 80mS).

Output Cable Resistance Compensation

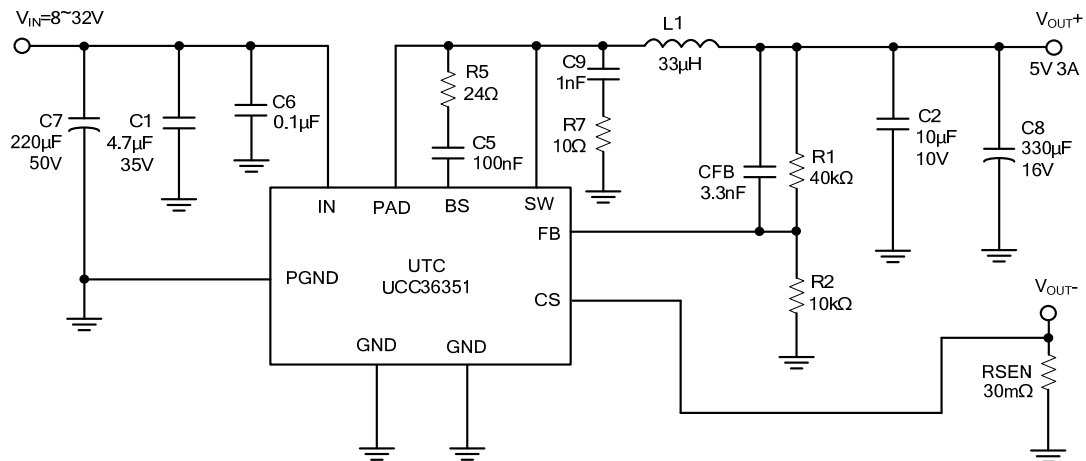
To compensate for resistive voltage drop across the charger's output cable, the UTC **UCC36351** integrates a simple, user-programmable cable voltage drop compensation using the impedance at the FB pin. Use the curve in Figure 1 to choose the proper feedback resistance values for cable compensation. $R1$ is the high side resistor of voltage divider.

$$V_{OUT} = V_{FB} \times (1 + R1/R2) + R1 \times I_{FB}(4\mu A)$$

TYPICAL APPLICATION CIRCUIT



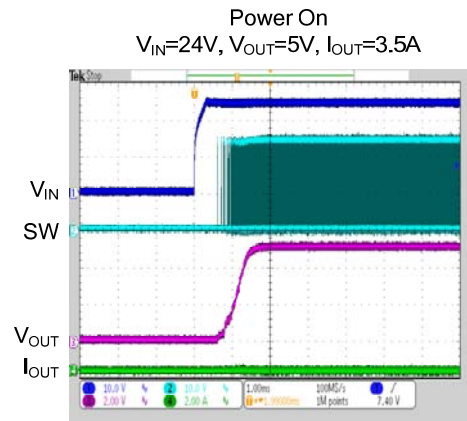
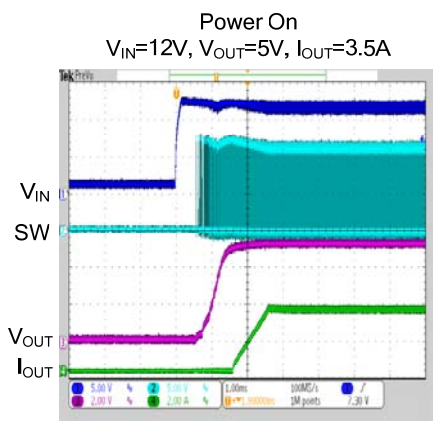
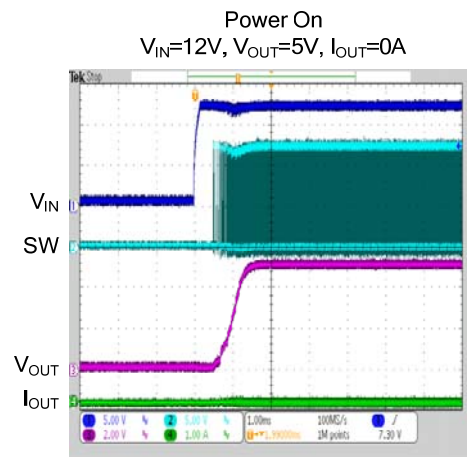
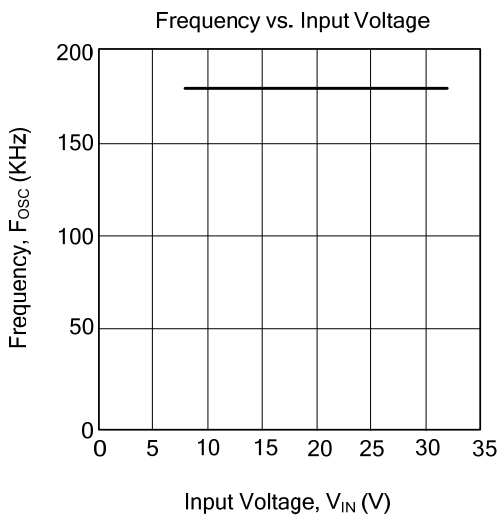
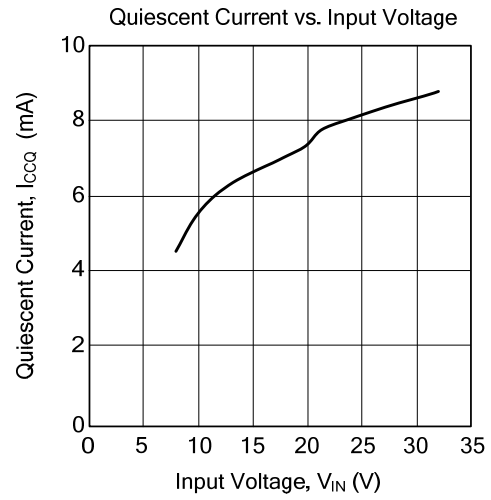
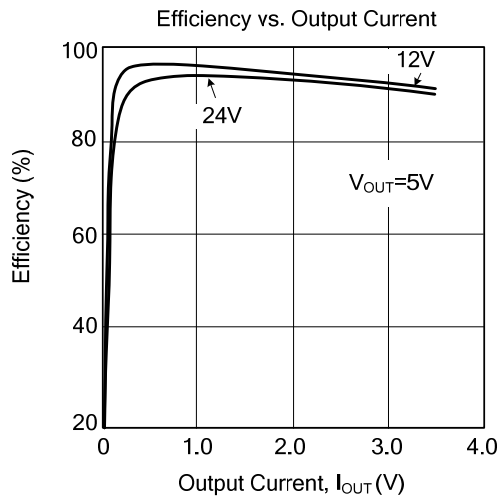
Note: $V_{OUT} = V_{FB} \times (1 + R1/R2)$, $V_{FB} = 1.00V$, $R2$ suggest $10k \sim 40k\Omega$, $I_{SEN} = 3.3A$ ($I_{SEN} = V_{CS} (0.1V) / R_{SEN} (30m\Omega)$)



Note: C9, R7 for EMI

Cable Compensation at Various Resistor Divider Values

■ TYPICAL CHARACTERISTICS



■ TYPICAL CHARACTERISTICS (Cont.)



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